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Problem Definition:

This project inspects three communication schemes (BPSK, OOK, and QPSK) in order to determine the achievable BER for each scheme under different noise conditions. The theoretical BER for each SNR value will be compared to the simulated results (Monte-Carlo Simulation) for each scheme. In the end, all schemes will be compared over the same range of SNR values.

Binary Phase Shift Keying (BPSK):

For BPSK the message, $\mathbf{X} \in [-1, 1]$ and $\mathbf{N} \sim \mathcal{N}(0, \sigma_n^2)$. This means that, assuming equal prior probabilities for each symbol, the Bayesian Decision Rule gives threshold at x = 0. The signal is given by,

$$H_0: \mathbf{X} = \mathbf{N},\tag{1}$$

$$H_1: \mathbf{Y} = \mathbf{S} + \mathbf{N}. \tag{2}$$

In this case, the theoretical value for the BER is given by,

$$P_e = Q\left(\sqrt{2\gamma_b}\right) \tag{3}$$

where Q(.) is the q-function, and γ_b is the SNR per bit. In order to test this theoretical value, a Monte-Carlo simulation is used to give an experimental result. Figure 1 represents the results of the Monte-Carlo solution, and Listing 1 below provides the code to generate these results.

```
Listing 1: Matlab Code – BPSK Simulation
```

```
1 %% BPSK Monte-Carlo Simulation
2 snr_max = 9.0; %dB
3 snr_min = 2.0; %dB
4 snr_step = 0.2;
5
6 snr_range = snr_min:snr_step:snr_max;
7
8 thresh_BPSK = 0;
9
10 num_transmissions = 1e6;
11
12 % set up figure
```

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```
13 figure
14 hold all
15 grid on
16
    grid minor
17
18
    error theo = zeros(size(snr range, 2), 1);
19
    \operatorname{error}_{\operatorname{sim}} = \operatorname{zeros}(\operatorname{size}(\operatorname{snr}_{\operatorname{range}}, 2), 1);
    for i = 1: size (snr range, 2)
20
21
         received message = zeros(num_transmissions, 1);
22
         \% calculate variance from SNR
23
         N 0 = 1/snr range(i);%1;% noise variance
24
25
         sigma sq = N 0/2;
26
27
         % generate random signal of 1's and 0's (P(0)=P(1))
28
         bits = randi(\begin{bmatrix} 0 & 1 \end{bmatrix}, num transmissions, 1);
29
30
         \% for BPSK, the bits should be -1 and +1, so 0 -> -1
         X = bits;
31
32
         X(X == 0) = -1;
33
34
         \% generate random gaussian noise
35
         Z = sqrt(sigma sq).*randn(size(X, 1), 1);
36
37
         % add noise to signal
         \mathbf{Y} = \mathbf{X} + \mathbf{Z};
38
39
40
         % check signal against threshold
41
         ind = find (Y> thresh BPSK);
         received message(ind) = 1;
42
43
         % get simulation bit error rate (BER)
44
45
         errors = received message (received message \tilde{} = bits);
46
         error sim(i) = size(errors, 1) / size(X, 1);
47
48
         % calculate theoretical BER
49
         \operatorname{error\_theo}(i) = \operatorname{qfunc}(\operatorname{sqrt}(2*\operatorname{snr\_range}(i)));
50
    end
51
    plot(snr_range, log10(error_sim), 'ko')
52
53
    plot(snr range, log10(error theo), 'k--')
    title ('Detector Performance with Varying SNR (BPSK)')
54
55 xlabel('SNR [dB]')
56
   ylabel('BER')
   legend('Simulated BER', 'Theoretical BER')
57
```



Figure 1: This figure depicts the results of a Monte-Carlo simulation ($1e^6$ iterations) to determine the relationship between SNR and BER the BPSK communication scheme. It is clear from the graph that the theoretical BER is comparable to the simulated BER.

On/Off Keying (OOK):

For the OOK communication scheme, the initial setup is the same as for BPSK. However, the only difference is that $\mathbf{X} \in [0, 1]$. This means that the threshold for detection is at x = 0.5, and that there is less 'space' between the on and off state. This will result in a degradation of BER performance as compared with BPSK, as can be seen in 2. The theoretical BER is given by,

$$P_e = Q\left(\sqrt{\gamma_b}\right),\tag{4}$$

The code to simulate the OOK setup is given in Listing 2 below.

```
Listing 2: Matlab Code - OOK Simulation
```

```
\operatorname{snr} \max = 12.0; \ \% dB
 1
 2
    \operatorname{snr} \operatorname{min} = 5.0; \ \% dB
 3
    snr step = 0.2;
 4
 5
    range = snr min: snr step: snr max;
 6
 7
    thresh OOK = 0.5;
 8
 9
     num transmissions = 1e6;
10
```

```
11 % set up figure
12 figure
13 hold all
14 grid on
15
   grid minor
16
17
    error_theo = zeros(size(range, 2), 1);
    for i = 1: size (range, 2)
18
19
        received message = zeros(num_transmissions, 1);
20
        \% calculate variance from SNR
21
22
        N = 0.5 / range(i);\% noise variance
23
        sigma sq = N 0/2;
24
25
        % thresh = some equation based on noise variance;
        % generate random signal of 1's and 0's (P(0)=P(1))
26
        bits = randi(\begin{bmatrix} 0 & 1 \end{bmatrix}, num transmissions, 1);
27
28
29
        \% for OOK, the bits should be 0 and +1
30
        X = bits;
31
32
        \% generate random gaussian noise
33
        Z = sqrt(sigma sq).*randn(size(X, 1), 1);
34
35
        % add noise to signal
36
        \mathbf{Y} = \mathbf{X} + \mathbf{Z};
37
38
        % check signal against threshold
39
        ind = find(Y > thresh OOK);
        received message(ind) = 1;
40
41
42
        % get simulation bit error rate (BER)
43
        errors = received message (received message \tilde{} = bits);
44
        error sim(i) = size(errors, 1) / size(X, 1);
45
46
        % calculate theoretical BER
47
        error theo(i) = qfunc(sqrt(range(i)));
48
   end
49
50
   plot(range, log10(error sim), 'ko')
51
   plot(range, log10(error theo), 'k--')
52 title ('Detector Performance with Varying SNR (OOK)')
53 xlabel('SNR [dB]')
54 ylabel('log(BER)')
55 legend('Simulated BER', 'Theoretical BER')
```



Figure 2: This figure depicts the effect of SNR on the BER for the OOK communication scheme. In this case, it is shown that the theoretical results agrees with the results of a Monte-Carlo simulation ($1e^6$ iterations)

Quadrature Phase-Shift Keying (QPSK):

In this section, the communication scheme will be identical to that of BPSK except that $\mathbf{X} \in \mathbb{C}$. This means that it is possible to send two bits of information per transmission (a real part and an imaginary part). Theoretically, the bit error rate is the same as BPSK, but this type modulation can allow for twice as much information to be sent. In order to test this, a Monte-Carlo simulation was done just as with BPSK and OOK. In this case, the noise added to this system was $N \sim \mathcal{CN}(0, \sigma_n^2)$. The results are given by Figure 3 and 4, and the code to generate them is given below in Listing 3. Figure 4 illustrates the data that the receiver would be processing in a noisy QPSK system.

Listing 3:	Matlab	Code -	OOK	Simulation
------------	--------	--------	-----	------------

```
%% QPSK Monte-Carlo Simulation
1
2
  %
  %
3
4
  %
      10 x
                  x 00
                          (odd bit, even bit)
5
  %
6
  %
                          real part (I channel)
7
  %
8
  %
                  x 01
      11 x
  %
9
```

```
10
11
   \operatorname{snr} \max = 9.0; \ \% dB
   \operatorname{snr} min = 2.0; %dB
12
13
   snr step = 0.2;
14
15
   snr range = snr min: snr step: snr max;
16
17
    num transmissions = 1e6;
18
19 % set up figure
20 figure
21 hold all
22
   grid on
23
    grid minor
24
25
   len = 1/sqrt(2);
   theta = zeros(num transmissions, 1);
26
27
   error theo = zeros(size(snr range, 2), 1);
    single errors = zeros(size(snr range, 1), 1);
28
29
    double_errors = zeros(size(snr_range, 1), 1);
    for i = 1: size(snr_range, 2)
30
31
        received message = zeros(num transmissions, 1);
32
         received = zeros(num transmissions, 1);
33
        \% calculate variance from SNR
        N 0 = 1/snr range(i);% noise variance
34
35
        sigma sq = N 0/4;
36
        \begin { figure } [h!]
    \langle begin \{ minipage \} [b] \{ 1.0 \setminus linewidth \}
37
38
      \ centering
      \centerline{\includegraphics[width=15cm, height=10cm]{EE526 Project2 OOK}
39
          .png\}
40 % \forall vspace \{2.0cm\}
41 \end{minipage}
42 %
43 \langle caption \{ \}
44 \label{fig:ook ber}
45
    \langle end \{ figure \} \rangle
        % generate random signal of 0's, 1's, 2's, 3's of equal probability
46
47
        % that represent the 4 positions for symbos in QPSK
48
        bits = randi (\begin{bmatrix} 0 & 3 \end{bmatrix}, num_transmissions, 1);
49
        % for QPSK, the bits should be
50
        X = bits;
51
52
53
        X(X == 0) = len + len*1i; \% quadrant
                                                   1 = 00
54
        X(X = 1) = -len + len*1i; \% qudrant 2 = 10
55
        X(X = 2) = -len + -len*1i;% quadrant 3 = 11
```

```
X(X = 3) = len - len*1i;\% quadrant
56
                                                      4 = 01
57
 58
         % generate random complex gaussian noise
         Z = sqrt(sigma sq).*randn(size(X, 1), 1) + sqrt(sigma sq).*randn(size(\leftrightarrow
 59
             X, 1), 1)*1i;
 60
 61
         % add noise to signal
 62
         \mathbf{Y} = \mathbf{X} + \mathbf{Z};
 63
 64
         % check signal against threshold
 65
          for j = 1:num transmissions
 66
 67
              rec bits = [real(Y(j)) imag(Y(j))];
             if(real(Y(j)) > 0 \&\& imag(Y(j)) > 0)
 68
                 % quadrant 1
 69
                  received _{message(j)} = 0;
 70
 71
             elseif (real(Y(j)) < 0 \&\& imag(Y(j)) > 0 )
 72
                  % quadrant 2
 73
                  received message(j) = 1;
 74
             elseif(real(Y(j)) < 0 \&\& imag(Y(j)) < 0)
 75
                  % quadrant 3
 76
                  received message(j) = 2;
 77
             else
 78
                  %quadrant 4
 79
                  received message(j) = 3;
 80
             end
81
         end
82
 83
          received (received message = 0) = len + len*1i; % quadrant 1 = 00
          received (received message == 1) = -\text{len} + \text{len}*1\text{i}; % qudrant 2 = 10
84
          received (received message == 2) = -len + -len*1i;% quadrant 3 = 11
 85
86
          received (received message = 3) = len - len*1i;% quadrant
                                                                                4 = 01
 87
 88
         % get simulation bit error rate (BER)
 89
          diff = abs(received - X);
90
91
          double errors(i) = size(diff(diff > 1.5), 1);
          single errors(i) = size(diff(diff > 1 & diff < 1.5), 1);
92
          \operatorname{error\_sim}(i) = (\operatorname{single\_errors}(i) + 2*\operatorname{double\_errors}(i))/(2*\operatorname{size}(X,1));
 93
94
95
         %calculate theoretical BER
          \operatorname{error\_theo}(i) = \operatorname{qfunc}(\operatorname{sqrt}(2*\operatorname{snr\_range}(i)));
96
97
     end
98
99 %plot
     plot(snr range, log10(error sim), 'ko')
100
     plot(snr range, \log 10(error theo), 'k--')
101
```

```
102 title('Detector Performance with Varying SNR (QPSK)')
```

- 103 xlabel('SNR [dB]')
- 104 ylabel('log(BER)')
- 105 legend('Simulated BER', 'Theoretical BER')



Figure 3: This figure represents the results of a Monte-Carlo simulation ($1e^6$ iterations) to verify the theoretical BER based on a given SNR value. As can be seen by the graph, the theoretical results agree with the experimental results.



Figure 4: This figure represents the noisy data that is received by the detector. The distribution around the true signal (red circles) is due to the noise added by the system. Here it can be seen how too much of an error will confuse the detector and cause an error.

Summary and Conclusions:

In closing, the most efficient scheme is QPSK. This is due to the fact that it can transmit twice as much information as OOK and BPSK, while achieving the same BER as BPSK (which is better than OOK). The BER for all schemes over a range of SNR values is given by Figure 5. The figure clearly depicts that the OOK has the worst performance (in terms of transfer rate vs. BER), while QPSK and BPSK are the same.



Figure 5: This figure represents the relationship between SNR and BER. Here it is clear that the BPSK and QPSK schemes outperform the OOK scheme in terms of BER. Also, it is clear that the BPSK and QPSK schemes have the same performance per bit.